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FINAL REPORT PRELIMINARY DESIGN OF A 15 M DIAMETER MECHANICALLY SCANNED DEPLOYABLE OFFSET ANTENNA

CONTRACT NAS5-26495

Submitted To

NATIONAL AERONAUTICS AND SPACE ADMINISTRATION

GODDARD SPACE FLIGHT CENTER

Greenbelt Road

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PREFACE

This report documents the results of Contract NASS-26495. The preliminary design of a 15 meter diameter, mechanically scanned, offset rotating, fed parabolic reflector antenna system is reported and the results of preliminary performance, structural and thermal analyses are presented.

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1.0 INTRODUCTION

A small 0.2 man year study has been accomplished on the preliminary design of a 15 meter diameter offset antenna system for application in a mechanically scanned high efficiency multi-frequency radiometer has been accomplished. The design activity results, presented in Section 2.0, and the analytical activity, presented in Section 3.0, succeeded in developing a mechanical design which is compatible with the design requirements. That is the mechanical design will provide an extremely high main beam efficiency. The effects of thermal distortion and mechanical scan are not system drivers.

The basic electrical analysis uncovered a substantial concern. This is the main beam efficiency effect of off-axis scan of the low frequency (5.1 GHz and below) beams. As indicated in the report, a simple, ideal circular feed aperture when scanned off-axis is not able to provide, for an ideal mechanical system, the desired 90% main beam efficiency. The results indicate that for the 5.1 GHz frequency the scanned beam efficiency will be no greater than 0.83. This can be improved by employing a larger under-illuminated long focal length aperture for the low frequencies similar to the 11 GHz application or by employing a complex shared aperture shaping feed and beam forming network. The more complex feed would bring associated network losses while the larger long focal length aperture would add significant mechanical impacts.

The results indicate that the desired on-axis main beam efficiency requirement of 90% is probably obtainable for all frequencies as is the 90% scanned beam efficiency at 11 GHz with the significantly longer focal length system.

2.0 SYSTEM DESCRIPTION

2.1 REQUIREMENTS

Established system ground rules require a 15 meter effective circular aperture offset parabolic deployable antenna mechanically scanned at 6 rpm in a conical pattern having a 35° half angle about NADIR. It will be operated on a spacecraft in a 700 km 12:00 sun-synchronous orbit. The antenna shall operate at frequencies of 1.414, 4.3, 5.1 and 11 GHz. The feed shall consist of a 10 horn array for operation at 4.3 GHz and a 5 horn array for operation at 1.4 GHz. The antenna shall be dynamically balanced at the required spin rate. The complete antenna system shall be capable of being launched on a single Space Shuttle flight in a launch volume not greater than 4 m diameter by 7 m long. The stowed configuration structural resonance frequency shall be > 25 Hz, and the scanning resonance frequency shall be > 12 Hz.

2.2 ANTENNA DESCRIPTION

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The antenna system chosen to meet these requirements is a 15 m diameter parabolic wrap rib reflector offset fed by a multiple horn array. Stowage requirements are met by mounting the feed and reflector on extendible triangular trusses, and the antenna system is dynamically balanced by counterweights mounted on extendible booms. The complete antenna assembly is depicted in Figures 2.2-1 and 2.2-2, and key ground design features are summarized in Table II-1.

The stowed configuration is shown in Figure 2.2-2 and the deployment sequence is described by Figure 2.2-3. Deployment is accomplished by first rotating the reflector mast cannister as shown in Step b. The reflector

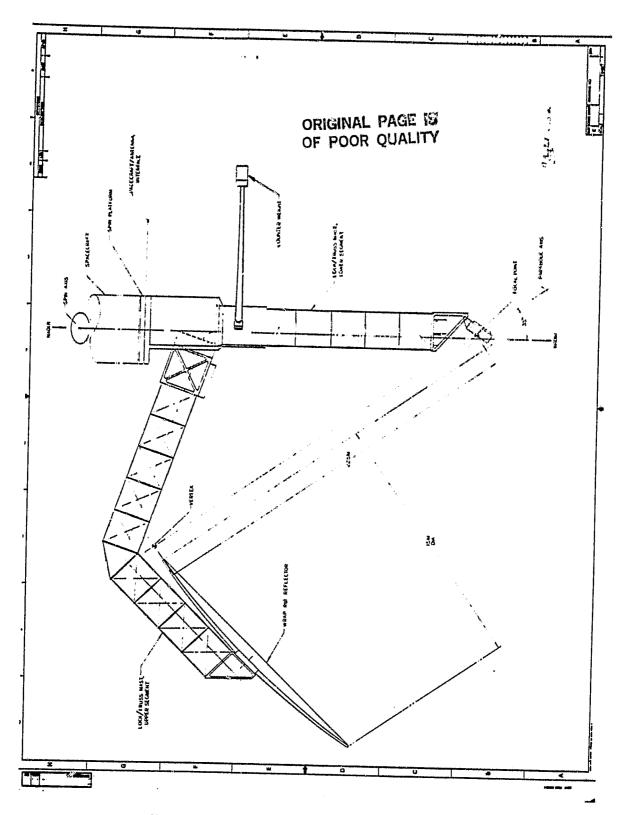


Figure 2.2-1 MSDA Configuration

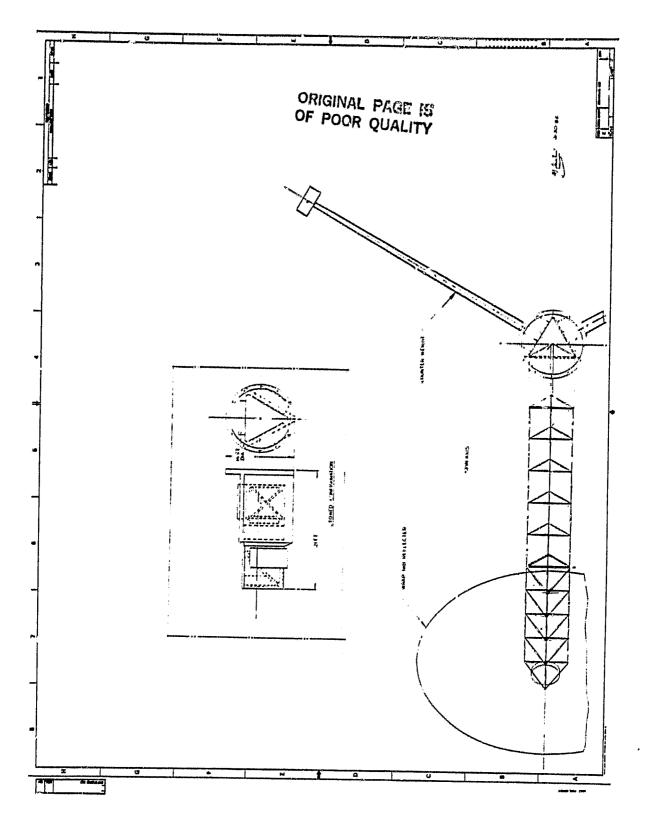


Figure 2.2-2 MSDA Configuration

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Table II-1 Antenna System Configuration

PARAMETER	DESIGN	REQUIREMENT
APERTURE DIA (m)	15	1.5
ARRANGEMENT	OFFSET PARABOLIC REFLECTOR	OFFSET PARABOLIC REFLECTOR
SCANNING	MECHANICAL 70° CONE	MECHANICAL 70° CONE
BEAM EFFICIENCY	91% (On Axis)	90%
FREQUENCY (GHz) BANDWIDTH (MHz) NO. OF BEAMS IN TRACK	1.414 28 3	1.414 28 3
FREQUENCY (GHz) BANDWIDTH (MHz) NO. OF BEAMS IN TRACK	4.3 200 10	4.3 200 10
FREQUENCY (GHz) BANDWIDTH (MHz) NO. OF BEAMS IN TRACK	5.1 100 12	5.1 100 12
FREQUENCY (GHz) BANDWIDTH (MHz) NO. OF BEAMS IN TRACK	11 100 10	11 100 10
NO. OF REFLECTOR RIBS	30	-
MASS	1045 Kg	-
INERTIA		
IxY	241,000 Kg-m ²	-
$\mathtt{I}_{\mathtt{xZ}}$	241,000 Kg-m ²	-
IxY	253,000 Kg-m ²	-
ANGULAR MOMENTUM	158,965 <u>kg-m²</u> sec	· -

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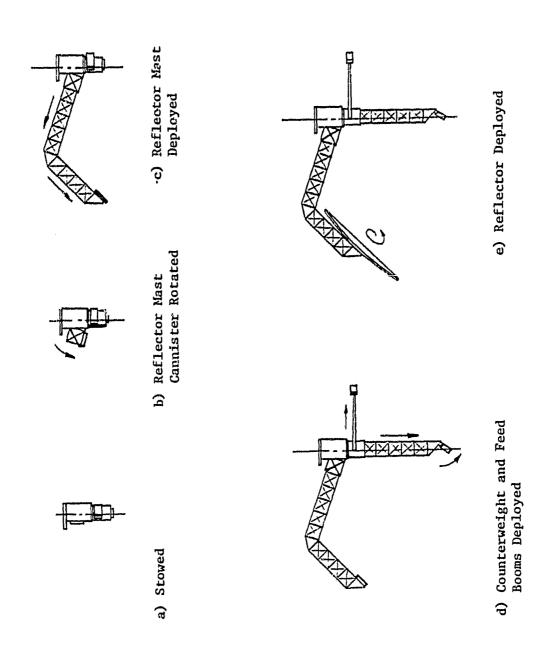


Figure 2.2-3 MSDA Deployment Sequence

mast is then extended as shown in Step c. The counterweight booms are extended next, the feed mast is extended, and the feed assembly rotated to its deployed position. Finally, the 15 meter reflector is unfurled to complete the system deployment. The MSDA is then ready for spin up to its 6 rpm operating rotation speed.

2.3 REFLECTOR DESCRIPTION

h

The reflector concept chosen is based on flight proven designs, some of which are depicted in Figure 2.3-1.

The wrap rib concept first conceived in 1963 was demonstrated at 9.2 meters and 8.3 GHz in 1974 aboard the ATS-6. LMSC recognized at about the same time the future needs for larger apertures operating at higher frequencies. Development of graphite epoxy technologies, mesh surface technology, reflector motor deployment technology, and high precision, repeatable reflector deployment mechanism technologies were initiated and pursued.

A product of these programs was the realization of the critical elements of two designs for 15 meter reflectors; both using G/E technology. These designs (Figure 2.3-2 and 2.3-3), while both generically wrap rib reflectors, differ in the technology development they were intended to provide.

The free deployment reflector, Figure 2.3-2, designed to operate at 12.5 GHz, was developed to provide a very lightweight medium diameter (less than 20 m) system. The hardware depicted in Figure 2.3-3 was designed to ascertain the limits to which G/E technology could be extended. The requirements were to develop the highest stiffness to weight ratio achievable. This reflector, originally intended to support a Shuttle Experimental Mission, was a predecessor model for extremely large apertures. This hardware became the basis for the design of the NASA Large Space Systems Technology 55 meter offset reflector (see Figure 2.3-4) currently

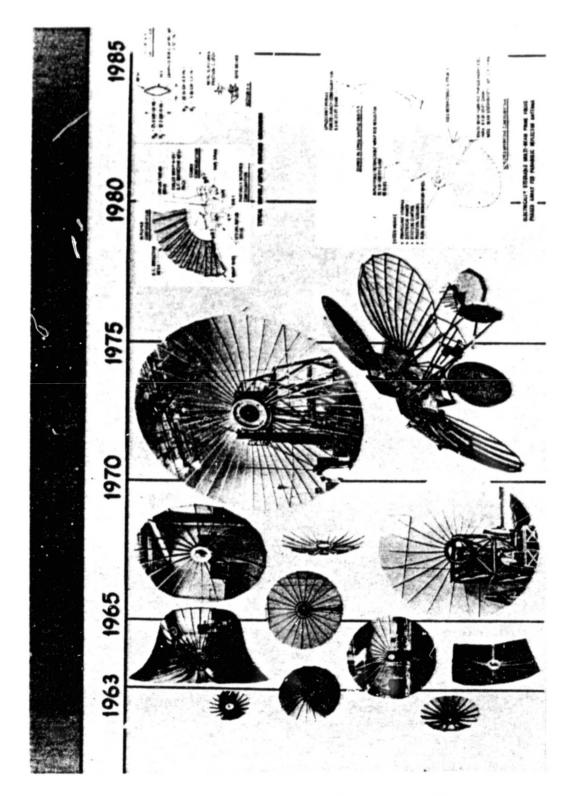


Figure 2.3-1 Wrap Rib Reflector Development History

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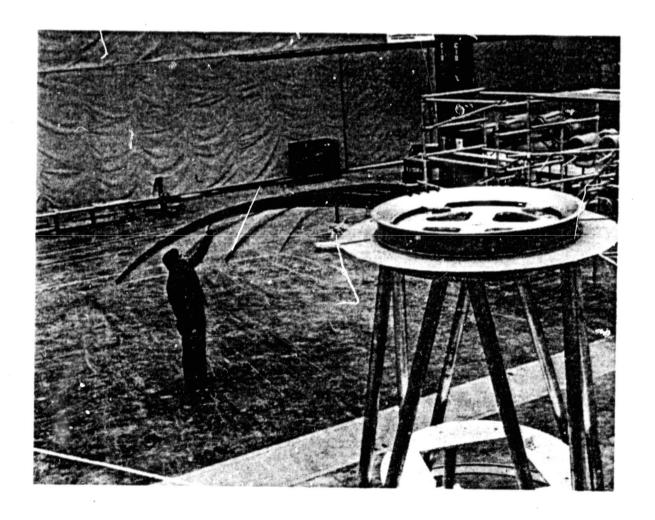


Figure 2.3-2 15 m Free Deployment Reflector Segment

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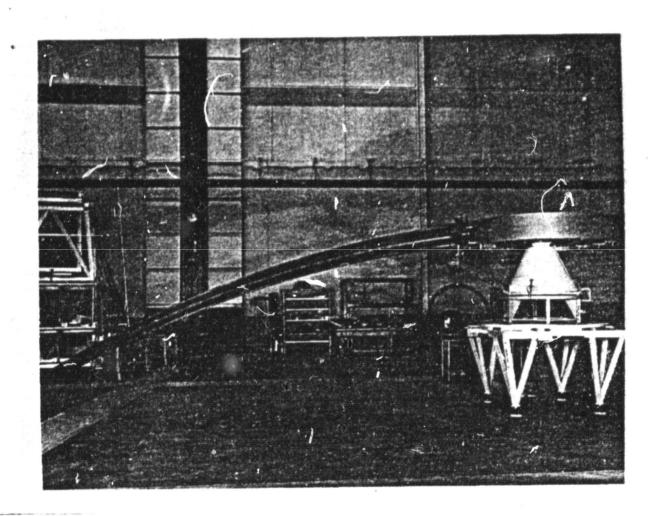


Figure 2.3-3 15 m Motor Deployed Reflector Rib

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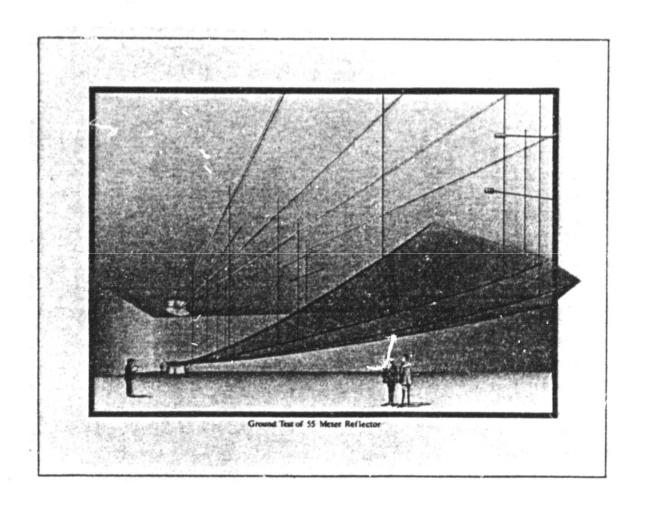


Figure 2.3-4 55 Meter Reflector Segment

being developed at LMSC under the auspicies of the Jet Propulsion Laboratory (JPL) and funded by the NASA Langley Research Center. The analysis techniques, design approaches, tooling requirements, and fabrication techniques for this reflector are all directly applicable to the MSDA reflector.

The wrap rib reflector consists of a number (variable) of radial ribs or beams which are cantilevered from a central hub structure. In an offset reflector, this hub is at the center of the offset section. Each of the ribs is attached to this hub through hinges. This radial spoke system provides the support for the reflective surface. The ribs are fabricated with appropriate contours to form the desired parabolic shape. Reflective, pie shaped, gores of a flexible membrane material (mesh) are attached directly to the ribs.

The rib cross section and material are chosen to permit the elastic buckling of the ribs. This allows them to be wrapped around the hub, spiral fashion, in the ascent (stowed) package configuration. In the stowing process, the ribs and attached reflective surface are rotated about the hinges until the ribs are tangent to the hub. After this rotation, the ribs are elastically buckled and wrapped into the receiving container.

The elastic energy stored in the wrapped ribs is sufficient to accomplish deployment of reflectors in this 15 m size range. A motor drive system controls the deployment speed allowing a slow, fully reversible deployment event.

Graphite epoxy was chosen as the rib material due to the inherently low coefficient of thermal expansion that can be achieved. The particular material composition and orientation selected is the Fiberite Company 0.250 mm. HMS/34 tape in a (0°/90°/90°/0°) laminate. The properties of these lamina materials and the resulting composite laminate are listed in Table II-2. The cross section designed for this rib is shown in Figure 2.3-5, and its key design parameters are shown in Table II-3.

Table II-2 Material Properties of Layup Configurations

PARAMETER	HMS/HMS/HMS
	(0/90 ₂ /0)
Young's Modulus (msi)	14.0
Shear Modulus (msi)	0.60
Thermal Coefficient of Expansion (x 10 ⁻⁶ /°F)	0.1
Ultimate Tensile Strength (ksi)	57.0
Ultimate Compressive Strength (ksi)	51.0
Thermal Conductivity (Btu/hr-ft-°F)	13.6

Table II-3 Rib Parameters*

PARAMETER	VALUE
Rib Type	Lenticular
Height (H) (Root)	178 mm
(tip)	50 mm
Width (W) (Root)	38 mm
(tip)	13 mm
Forming Radius (R)	68.7 mm
Length (L)	7.5 m
Thickness (T)	.5 mm
Hub Radius (r)	1.46 m

*See Figure 2.3.5 for physical description of rib

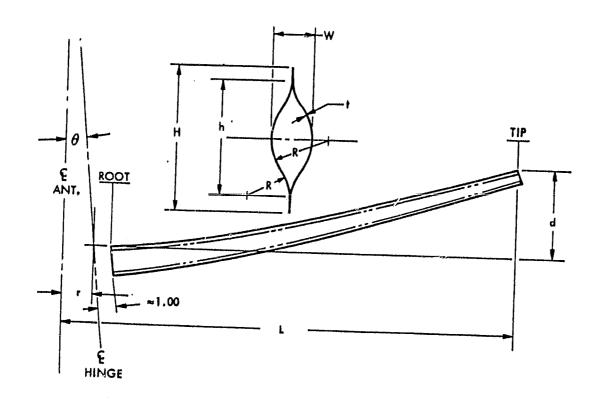


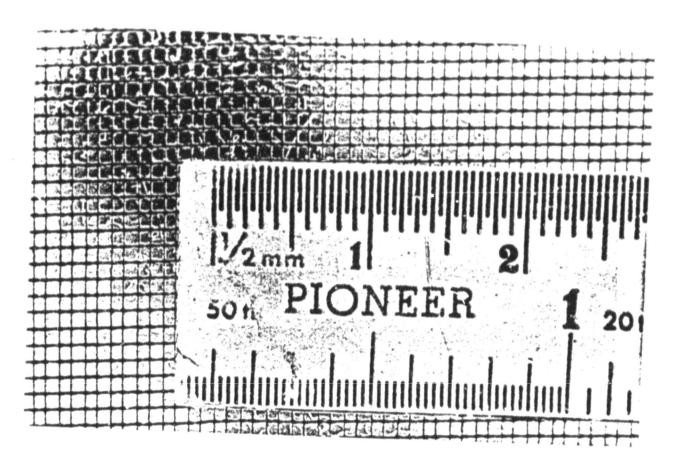
Figure 2.3-5 Lenticular Rib Geometry

Mesh attachment is provided for by the addition of hollow eyelets installed near the parabolic edge of the rib. The purpose of these eyelets is to protect the sewing thread from the chafing that would result from bare graphite epoxy holes.

The mesh chosen for the MSDA reflector (illustrated in Figure 2.3-6) is a Dacron fiber woven into a leno weave marquisette and manufactured by Travis Mills Corporation; its major characteristics are listed in Table II-4. The Dacron mesh is copper-coated and then has an outer protective silicone coating. The designation for this mesh is T635/Cu/6-1104. It is the same mesh used on the ATS-6 reflector, and its properties are fully reported in Document No. 5177259 titled "ATS F&G Parabolic Reflector Subsystem Antenna Mesh" dated January 15, 1974. LMSC has extensive experience with the T635 mesh and its use as a reflector material. Electrical performance of the mesh is discussed further in Section 3.2.2.

2.4 FEED MAST DESCRIPTION

The demand for large space platforms and antenna systems has identified the need for extremely long, stiff deployable booms. This need has prompted Lockheed Missiles and Space Company to develop a deployable space mast structure capable of repeatable precision deployments in the space environment without external aids. The structure, shown in Figure 2.4-1, is made up of longitudinal columns and cross braces which are doubly tapered graphite-epoxy tubes for maximum strength/weight ratio and stowing efficiency. The longitudinal members are hinged at the midpoint with a simple virtual hinge mechanism to provide a compact folding scheme. Small diameter tension rods serve as diagonal members. Pretensioning of these diagonals eliminates clearance in the longeron pivot bearings which ensures structural continuity and provides the majority of torstional stiffness. These features integrate into a truss system that exhibits a stiffness per unit length ratio and a stowage efficiency unsurpassed by any currently available design approaches.



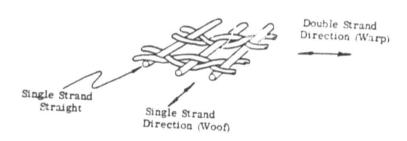


Figure 2.3-6 Woven Dacron Reflector Mesh

TABLE II-4

DACRON MESH SUBSTRATE

- 1. SOURCE Style 635, Travis Mills Corporation, New York, NY
- 2. TYPE Leno weave marquisette
- 3. YARN High tenacity dacron, 70 denier, 14 filaments

 Longitudinal direction: Twisted pairs of warp yarns,
 .98 pr/mm

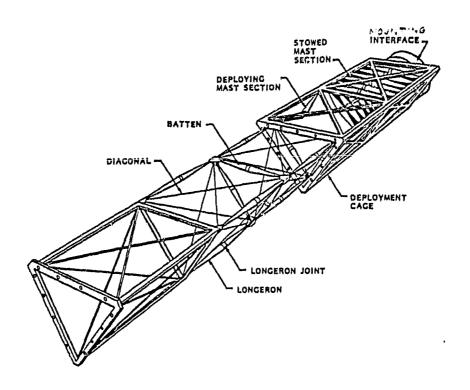
 Transverse direction: Single fill yarns, 1.18/mm

 Melamine sizing

 Filament diameter: 0.23 mm

Yarn cross section: $4.067 \times 10^{-5} \text{ mm}^2$

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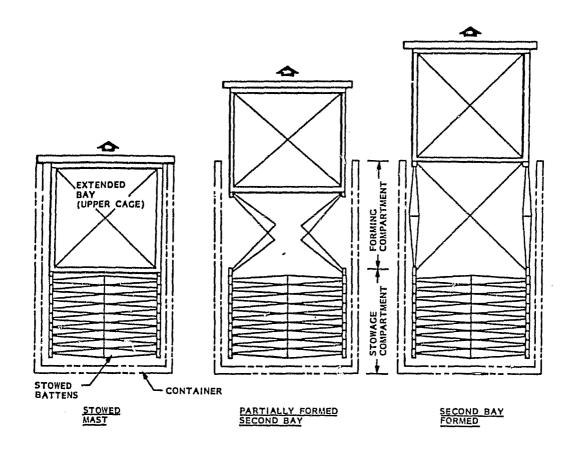
Figure 2.4-1 Feed Mast Components

The mast is deployed and retracted sequentially bay by bay, the extension sequence is illustrated in Figure 2.4-2. The stowed mast is held in deployment cages which are functionally divided into two compartments, one for handling the stowed mast and one for extending the bays. The stowed mast is slowly raised toward the forming compartment; the mechanism in the forming compartment lifts a single batten assembly and extends the longerons and diagonals of one bay until the bay is fully formed. This process is repeated until the complete mast is fully formed. These steps are reversed to retract the mast. A simplified schematic of the device is pictured in Figure 2.4-3 which illustrates the high/low speed drive arrangement and the function of the gear boxes, driver and belts. A significant design feature of the deployment mechanism is that during deployment loads are transmitted through the deployed mast sections into the upper deployment cage and around the deploying sections. This feature assures a predictable structural stiffness throughout the entire deployment operation.

The mast physical properties are detailed in Table II-5. The bay size was selected to provide the greatest cross section compatible with the STS Orbiter Cargo Bay stowage requirement. The other key design parameters are chosen to provide the highest overall deployed vibration frequency compatible with the known kinematic and system stowage constraints.

The mast design has been carried to construction of individual scale longeron elements to verify the longeron fold joint kinematics and beam strength.

The reflector support mast is of the same design as the feed support mast except that it contains one unique bay which forms the "knee joint" near the middle of the structure.



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Figure 2.4-2 Mast Structure Deployment Schematic

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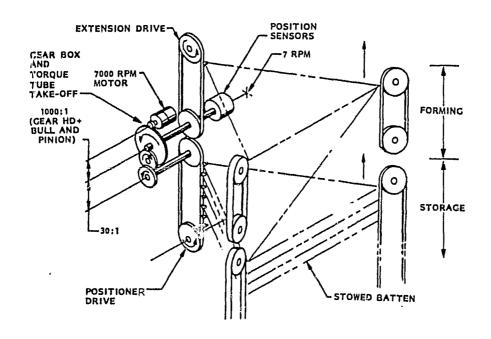


Figure 2.4-3 Mast Extension Drive Schematic

This bay is formed by deploying a single longeron and pivoting the mast about an axis passing through the other two corners of that bay, thus forming a single tetrahedular bay embedded within the normal mast. This arrangement allows the same mast deployment apparatus to deploy this bay.

2.5 COUNTERWEIGHT DESCRIPTION

The offset antenna requires a set of counterweights to dynamically balance the rotating system. Figures 2.5-1 and 2.5-2 shows families of mass/ offset curves which satisfy the dynamic balance requirement for the two possible spin axes giving a 35° scanning angle. These curves were obtained from a parametric computer optimizing routine assuming lumped masses of 270 kg at the feed and 185 kg at the reflector. Two masses of the sizes listed on the curves located at the points shown achieve dynamic balance for the reflector assembly. As can be seen from these figures arranging the spin axis as shown in Figure 2.5-1 is less sensitive to counterweight positioning tolerances. That spin orientation was therefore selected for the design. The masses are supported from the antenna system by means of extendible booms of the stem real type (see Figure 2.5-3). This design has been selected because of its simplicity, low stowed volume and extensive successful development and flight experience. Documentation of the bistem boom configuration and design parameter information can be found in the proceedings of the 2nd Aerospace Mechanisms Symposium of May 4 - 5, 1967, NASA Technical Memorandum TM 33-355.

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Table II-5 MAST PROPERTIES

PARAMETER		FEED MAST		REFLECTOR MAST
Length		13.3 m		21.3 m
Weight - Longerons Battens Diagonals	10.6 Kg	30.9 Kg	16.5 Kg 17.0 Kg 16.0 Kg	49.5 Kg
Bay Width		2.66 m .		2.66 m
Bay Aspect Ratio		1:1		1:1
Number of Bays		5		8
Longeron Diameter -	Maximum Minimum			10.0 cm 2.54 cm
Longeron Thickness		.79 mm		.79 mm
Material		Graphite Epoxy Thermal 50		Graphite Epoxy Thermal 50
Stiffness (Bending)		$289 \times 10^3 \frac{\text{N}}{\text{M}}$		$113 \times 10^3 \frac{\text{N}}{\text{M}}$
(Torsion)		$294 \times 10^3 \frac{N-M}{rad}$		$115 \times 10^3 \frac{N-M}{rad}$

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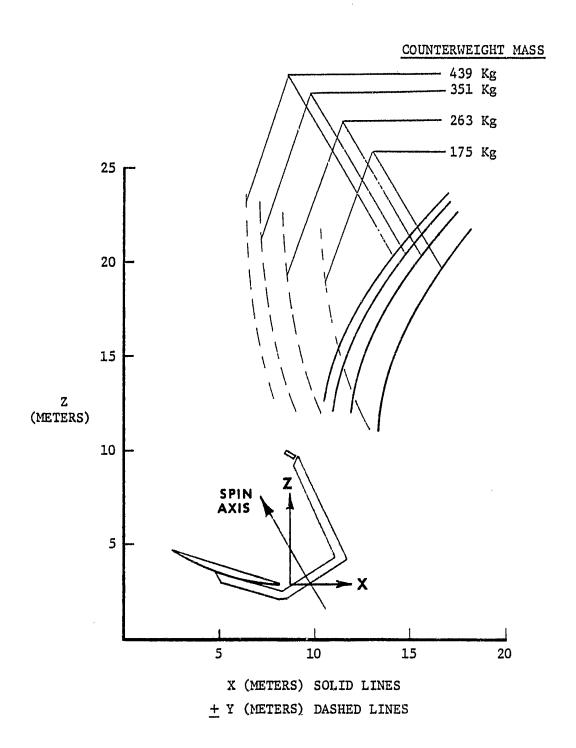


Figure 2.5-1 Counterweight Positions 25

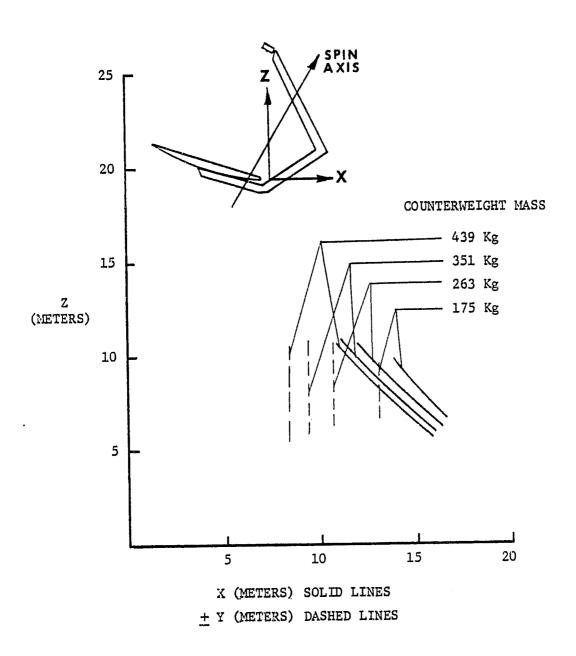


Figure 2.5-2 Counterweight Positions

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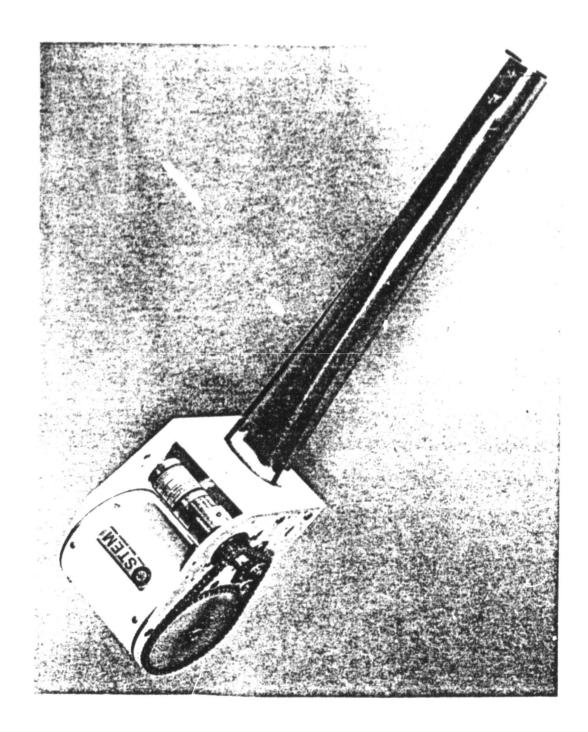


Figure 2.5-3 Extendible Reel (Bi-Stem) Boom

3.0 MAIN BEAM EFFICIENCY EVALUATION

The 15 m antenna system was analyzed to identify the individual error components which effect the overall main beam efficiency. These analyses which addressed electrical and mechanical error contributors are detailed in this section. In summary the maximum achievable main beam efficiency which can be expected for the off axis beam at frequencies of 5.1 GHz and lower is 0.83. This limitation results from the achievable performance of a simple feed aperture for the scanned beam case. The result should not be considered conservative since the feed model was ideal in that the pattern was perfectly axisymmetric which is difficult to achieve in hardware. The 11 GHz case is aided by the under illuminated aperture and the longer apparent focal length to yield a scanned beam efficiency of 0.89. This case illustrates the benefits which could be achieved at the low frequencies by employing multiple apertures for beam forming or extremely long focal lengths. The individual errors and overall performance developed for the antenna are displayed in Figure 3.0-1. These data are referenced to the appropriate sections which contain details of the analyses performed.

3.1 METHOD OF RF ANALYSIS

The secondary far field patterns of a parabolic reflector were obtained by using the efficient, versatile Fourier Bessel method just developed by Dr. Charles C. Hung of the RF/Antenna Systems, LMSC. The secondary field of an arbitrary reflector can be expressed in terms of physical-optics integral of the current on the reflector. There exist many techniques to evaluate, or compute, this physical optics integral. For scan cases all these previously developed techniques become essentially impractical from cost and computer time considerations. The Fourier Bessel

ERROR. CROSS POLARIZATION, CROSS POLARIZATION, SPILLOVER SCAN THERMAL* APPROXIMATION MANUFACTURING* ROTATION*	SECTION 3.2 3.3.1 3.3.4 3.3.5	0.9999 0.9999 0.9999 0.9999 0.9999	CGHZ SCANNED 0.9172 0.9378 0.9708 0.9991 0.9999	0.9237 0.9999 0.9987 0.9999	SCANNED 0.9237 0.9944 0.9999 0.9987 0.9975
TOTAL		0.8858	0.8307	0.8928	0.8878
*Based on Aperture Efficiency Calculation	ncy Calculati	no.			

Figure 3.0-1 15 m Antenna Main Beam Efficiency

method is very versatile and can be employed for any reflector shape such as a spherical reflector, reflectors with surface errors and those reflectors with projected aperture shapes that deviate from the familiar circular geometry, as for instance, a polygonal-shaped aperture.

In the Fourier Bessel method, the integrand of the physical optics integral for a reflector is first expanded in terms of the well known sinusoidal functions and the integration is carried out analytically with the aid of a shape function. This expansion of the integrand into sinusoidal functions does not depend on the reflector geometry and is accomplished via the well established Fast Fourier Transform (FFT) algorithm. The far field is then expressed in terms of a summation of a series of Airy functions for a circular projected reflector aperture and a series of tangent and exponential functions for a polygonal shaped projected reflector aperture. The convergence of this resultant series is very good even for wide angle scan cases and hence can be evaluated using very little computer time. The details of this Fourier-Bessel method to compute the secondary far field pattern will be published soon.

For scan beam cases, one important design information is the location of feed elements. This information can be obtained by using the focal region distribution of the reflector when it is operating in receiving mode and is illuminated by a plane wave incident from the intended direction of scan. By reciprocity theorem, the focal region distribution can be obtained by repeatedly computing the radiated far field of the same reflector for different feed positions in its focal region. With the efficient Fourier Bessel method, the focal region distribution can be generated without incurring high computer cost to provide the correct feed element location for the given scan angle.

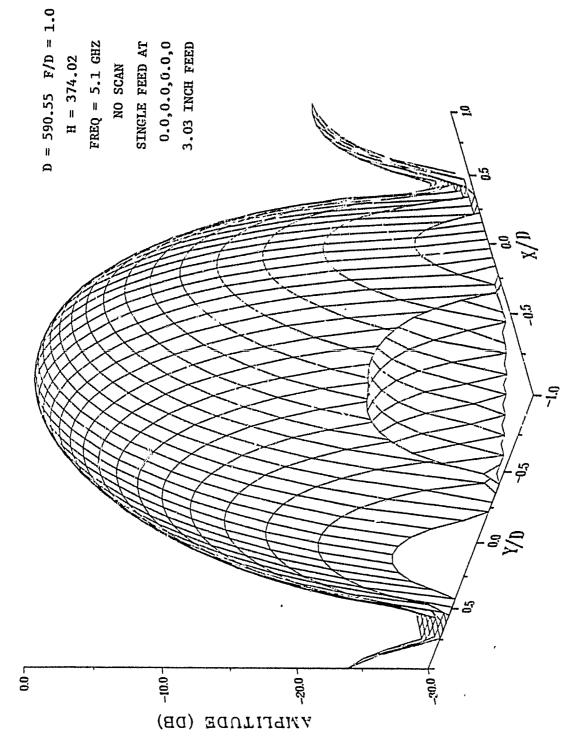
The main beam efficiency can be computed by integrating the main beam pattern and comparing to the total energy radiated. Again, with the efficient Fourier Bessel method of computing the far field pattern, it

is now possible to perform the pattern integration on the computer with a very reasonable computer time.

3.2 RF ANALYSIS

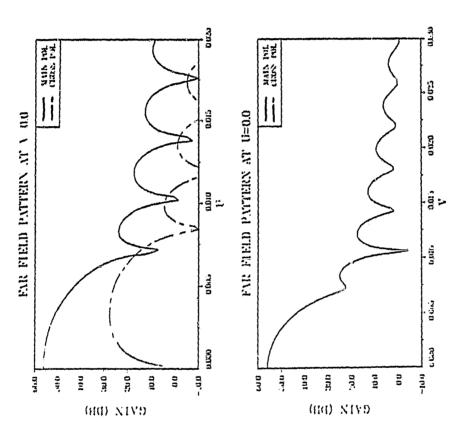
In order to determine the total main beam efficiency of the perfect offset antenna the far field patterns of the offset parabolic reflector were obtained. The feed element used in the analysis was a linearly polarized circular aperture with uniform distribution across the aperture, hence the feed pattern is an Airy function. The radius of this circular aperture was chosen to maximize the 5.1 GHz main beam efficiency of the on axis case for an F/D = 1.0 offset geometry. In this case the F/D represents the ratio of the actual system focal length to the projected circular aperture diameter perpendicular to the axis of the main beam. The main beam efficiency maximized at 91.72% with a feed aperture radius of 76.96 mm (3.03 in.). This maximum main beam efficiency feed diameter produces -10 dB power edge directed illumination. The resulting surface current distribution projected and circular aperture plane is shown in Figure 3.2-1 and the far field pattern in Figure 3.2-2. The calculated efficiency was obtained by pattern integration between the first nulls referenced to the total energy radiated by the feed. As a result this efficiency includes the effects of (1) aperture illumination, (2) cross polarization, and (3) spillover inherent in the calculation for the offset antenna.

After the on-axis case efficiency was optimized the case with the beam scanned 1.627 degrees or 5.5 beam widths off axis was calculated. For this case the direction of scan was from the focal point radially toward the outer edge of the reflector in the direction of the offset. The results are shown in Figure 3.2-3 with a calculated efficiency of 80.18%. This result indicated substantial scan impacts which are incompatible with the desired 90% efficiency requirement. The system F/D was then varied to evaluate the ability to reduce the scan effects to an



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Figure 3.2-1 Equivalent Surface Current, On-Axis, 5.1 GHz



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D = 590.55 F/D = 1.0

REFI ECTOR

H = 374.02

OFFSET PARALOLIC

FEED RADIUS = 3.03

(0.0,0.0,0.0)

FEED AT

FREQ = 5.1 GHZ

MAIN BEAM EFFICIENCY =

0.917

Figure 3.2-2 On-Axis Pattern, 5.1 GHz, F/D = 1.0

OFFSET PARABOLIC

CROSS POL

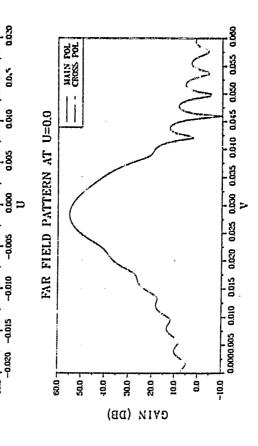
FAR FIELD PATTERN AT V=SIN(1.627)

60.0 50.0

30 200

ечім (рв)

-100



MAIN BEAM EFFICIENCY = 0.808

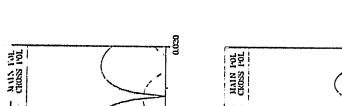
Figure 3.2-3 Scan Beam Pattern, 5.1 GHz, F/D = 1.0

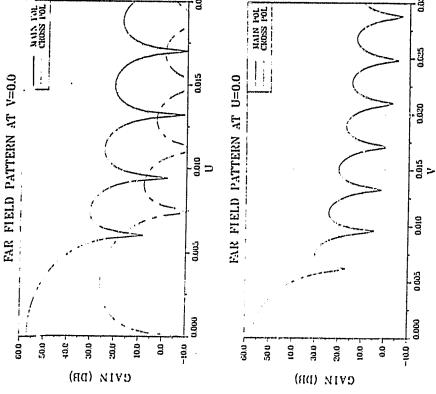
acceptable level. Since the problem was isolated to the effects of scan the feed was not re-optimized for the study of F/D efficiency variation. This results in the calculated efficiencies use being restricted to relative evaluation.

Figures 3.2-4, 3.2-5, 3.2-6 and 3.2-7 present the resulting far field patterns for the on axis and scanned cases for F/D = 1.25 and 1.5 respectively. These results indicate that the efficiency effect due to scan for the F/D = 1.25 case is 0.912 and the F/D = 1.5 is 0.9378. The on axis efficiencies for both cases can be increased to the 0.9172 calculated for the F/D = 1.0 case with re-optimized feed diameter so the achievable total efficiency for the scanned F/D = 1.25 system is 0.8365 and for the F/D = 1.5, 0.8553. Some additional improvement in these total efficiencies would result from re-calibration since the cross polarization effects are also reduced by the longer focal lengths and these reductions are not included without rigorous re-calculation.

At this point selection of an F/D = 1.5 is indicated as the necessary geometry to achieve a near 90% efficiency requirement. With this selection of F/D the under illuminated case was investigated. The feed function was adjusted by varying the feed radius to achieve an 0.35 degree beamwidth at 11.0 GHz. This was obtained with a radius of 81.28 mm (3.20 inches). The resulting aperture surface currents are presented in Figure 3.2-8. Both the on axis and 1.575 degrees scanned far field patterns were then calculated. These results are presented in Figures 3.2-9 and 3.2-10. The calculated main beam efficiencies are 0.9237 for the on axis beam and 0.9185 for the scanned beam.

The results of the pattern analysis thus indicate in summary that an F/D = 1.5 system is required to reduce the effects of scan losses on efficiency to acceptable levels. The calculations indicate a worst case main beam efficiency of 0.8553 at 5.1 GHz for the beam scanned 1.627 degrees (5.5 beamwidths) off axis and 0.9185 at 11 GHz for the beam scanned





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D = 590.55 F/D = 1.25

OFFSET PARABOLIC

REFLECTOR

FEED RADIUS = 3.03

H = 374.02

(0.0,0.0,0.0)

FEED AT

On-Axis Pattern, 5.1 GHz, F/D = 1.25Figure 3.2-4

MAIN BEAM EFFICIENCY =

020 03000,005 0.010 0.015 0.020 0.025 0.030 0.035 0.040 0.045 0.050 0.055 0.000 MAIN POL CROSS POL MAIN POL CROSS POL 0.015 FAR FIELD PATTERN AT V=SIN(1.627) 0.010 FAR FIELD PATTERN AT U=0.0 0.005 -0.005 0000 -0.020 -0.015 0.01 000 50.0 40.0 50.0 счім (вв) ечім (рв)

Figure 3.2-5 Scan Beam Pattern, 5.1 GHz, F/D = 1.25

MAIN BEAM EFFICIENCY =

0.8007

D = 590.55 F/D = 1.25

H = 374.02

REFLECTOR

OFFSET PARABOLIC

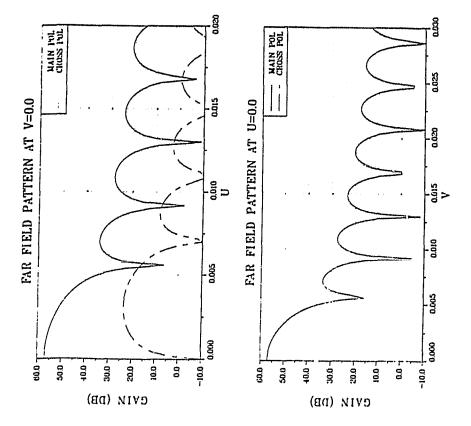
(0.0, -20.00, -11.00)

FEED RADIUS = 3.03

FEED AT

FREQ = 5.1 GHZ

ORIGINAL PAGE 19 OF POOR QUALITY



D = 590.55 F/D = 1.50

REFLECTOR

OFFSET PARABOLIC

FEED RADIUS = 3.03

(0.0,0.0,0.0)

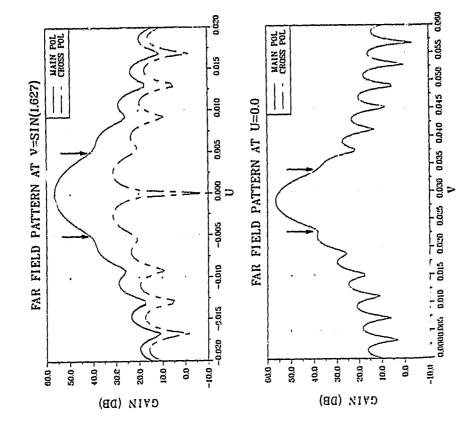
FEED AT

FREQ = 5.1 GHZ

H = 374.02

Figure 3.2-6 On-Axis Pattern, 5.1 GHz, F/D = 1.5

MAIN BEAM EFFICIENCY =



C

(II)

D = 590.55 F/D = 1.50

H = 374.02

OFFSET PARABOLIC

REFLECTOR

(0.0,-24.00,-11.50)

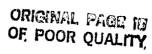
FEED RADIUS = 3.03

FEED AT

FREQ = 5.1 GHZ

Figure 3.2-7 Scan Beam Pattern, 5.1 GHz, F/D = 1.5

MAIN BEAM EFFICIENCY =



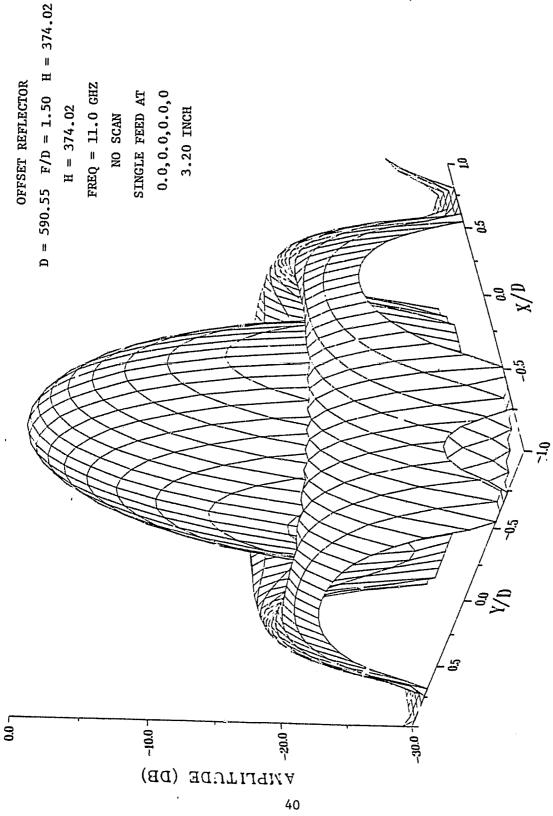
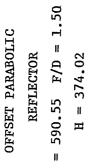
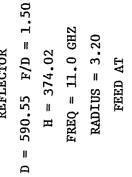


Figure 3.2-8 Equivalent Surface Current, On-Axis, 11 GHz





MAIN FOL CROSS POL

FAR FIELD PATTERN AT V=0.0

C0.00

50.0

30.0

CAIN (DB)

10.0

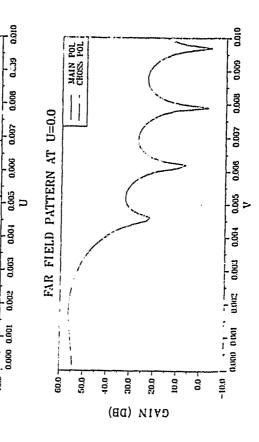
20.0

0.0

-10.0

4





MAIN BEAM EFFICIENCY =

Figure 3.2-9 On-Axis Pattern, 11 GHz, F/D = 1.5

OF POOK CONTENT

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D = 590.55 F/D = 1.50

H = 374.02

REFLECTOR

FREQ = 11.0 GHZ SINGLE UNIFORM

OFFSET PARABOLIC

(0.0,-23.50,-10.50)

FEED AT

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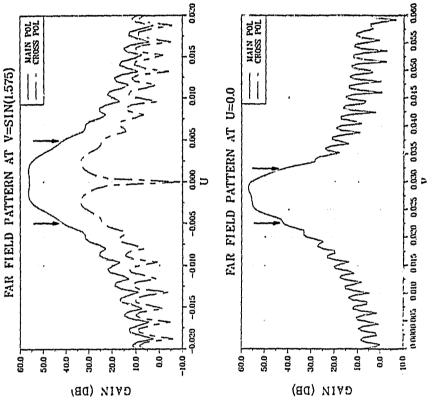


Figure 3.2-10 Scan Beam Pattern, 11 GHz, F/D = 1.5

MAIN BEAM EFFICIENCY =

1.575 degrees (4.5 beamwidths) off axis. These efficiencies include the effects of aperture illumination, cross polarization, spillover and off axis scan.

3.3 MECHANICAL ANALYSIS

The results of the electrical analysis indicate that in order to achieve a 90% main beam efficiency at 11 GHz the allowable mechanical error budget is 0.9799. As previously indicated 90% efficiency at 5.1 GHz will not be achievable. This must include the effects of surface approximation, mesh, thermal distortion and mechanical scan of the antenna system. The following is a description of the analysis performed to bound the effects on efficiency of the individual mechanical errors.

3.3.1 Thermal Distortions

The orbit thermal analysis performed on the 15 meter diameter reflector investigated the quarter points of each of three orbits considered. These were sun angles of 0, 30 and 80 degrees with respect to the orbit plane. At each of the quarter points one spacecraft revolution was investigated. Temperature information was obtained for each of the four points during the spacecraft spin. These data were provided for the subsequent distortion analyses. In addition the average temperature was obtained for the orbit positions investigated by averaging the four point spin temperatures.

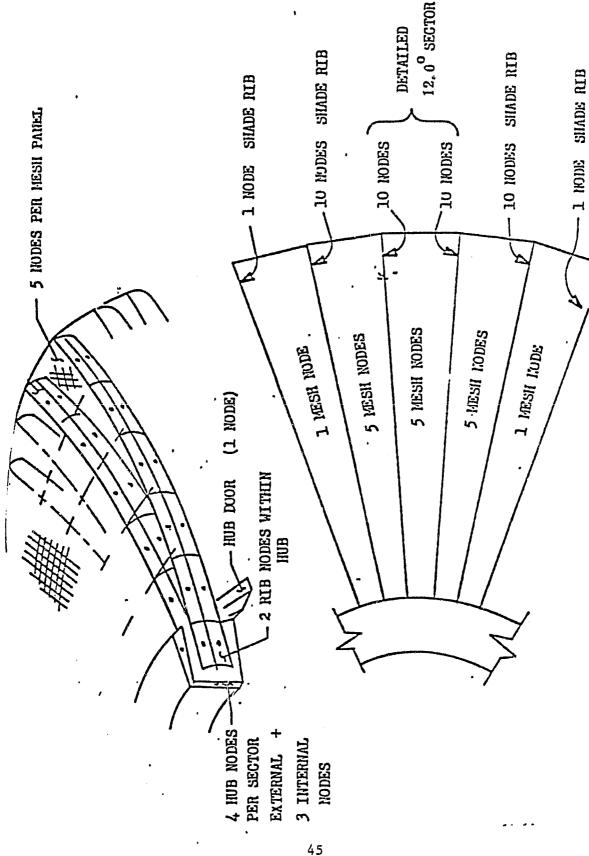
Based on previous study work on other programs it has been found that the rib depthwise temperature gradients, which have a primary effect on surface distortion, are proportional to the rib absolute temperature. Therefore, the colder the ribs run in orbit, the lower the temperature gradients and the lower the surface distortion. Since the antenna system was spinning and the solar arrays were assumed de-spun the effect

of the arrays is to cause the ribs, they repeatedly shadow during the spinning, to run colder. Since this would produce smaller rib gradients, it was felt that the arrays should be neglected thereby leading to conservative predictions for performance.

A thermal model previously developed was used to obtain the heat rates for each segment of the 15 meter reflector and was used to calculate the thermal load conditions. Figures 3.3-1 and 3.3-2 present the nodal. breakdown used in the segmented approach to the determination of heat rates. The temperature response for each sector was computed using the model shown in Figure 3.3-2 with four other sector heat rates combined with the sector of interest. The hub was modeled in detail as a typical hub structure. Figure 3.3-3 presents a cross section of a hub thermal model. The hub consisting of upper and lower rings and a vertical spacer is modeled in detail. The insulation around the hub is modeled with inner and outer nodes connected with a radiation resistance based upon an effective emittance factor. The outer surfaces of the inter-rib curtains and insulation are Kapton aluminized on the backside which results in an $\alpha = 0.4$ and $\epsilon = 0.6$. The inner surface of the curtain is aluminum with an $\varepsilon = 0.1$. The hub doors are anodized (DOW 17) which results in an $\alpha = 0.78$ and $\epsilon = 0.7$. The rib surface optical properties were assumed to be those of a surface coated rib with $\alpha = 0.20$ and $\varepsilon = 0.35$.

The thermal distortion analysis performed to obtain the distorted orbital surface description was performed on a finite element computer program SPAR. The program is based on the direct stiffness method of matrix structural analysis. This program handles 17 types of finite structural elements.

The reflector model employs quadrilateral constant thickness membrane elements to model the mesh. In-plane displacements vary linearly along the edges and within the element interior. Mesh stiffness properties



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Figure 3.3-1 Detailed Thermal Analysis Approach Nodal Breakdown

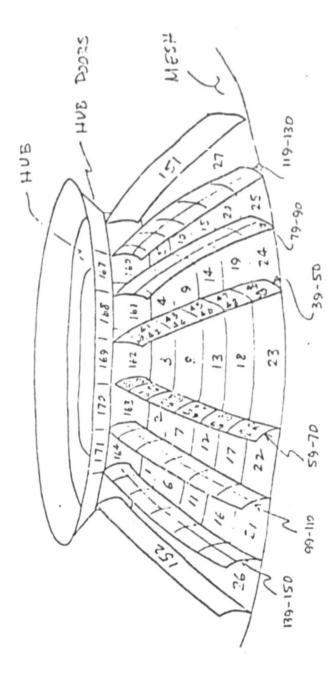


Figure 3.3-2 Nodal Breakdown

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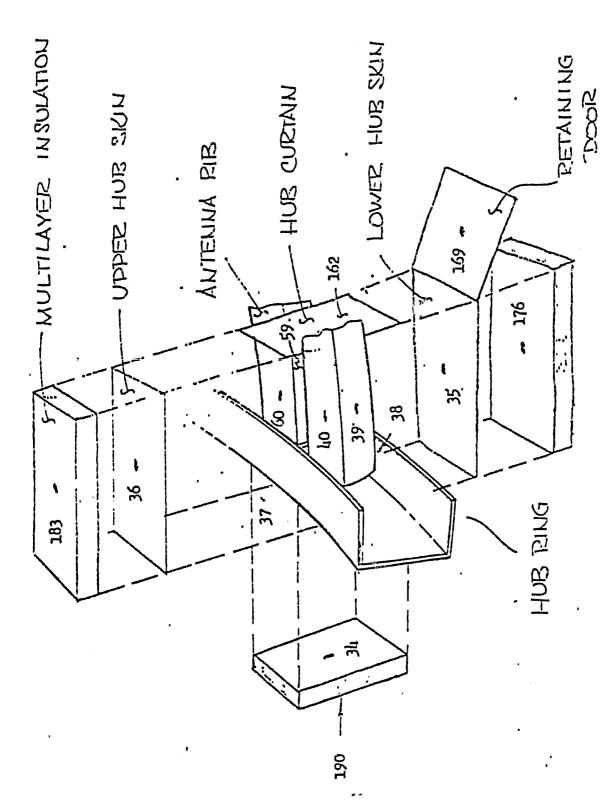


Figure 3.3-3 Hub Model

vary as a function of strain and temperature and the thermal coefficient of expansion varies as a function of temperature.

The ribs are modeled as beam elements with primary and secondary bending planes. The element line of mass and shear center are offset to allow for the proper description of the rib properties.

The hub is modeled with plate, ring and beam elements depending on the particular type of construction involved.

The loading conditions analyzed consist of the orbit thermal loads applied to the mesh elements, rib elements and hub elements. These loads are provided by absolute temperature changes as well as temperature gradients and differences through the rib and hub elements.

The results of the thermal distortion analysis in terms of surface weighted rms differences from the designed zero gravity wrap rib surface are presented in Figure 3.3-4. In addition, the efficiency effects so rigorous pattern calculation with the distorted surface was not pursued.

3.3.2 Mesh Effects

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Previous testing of meshes at LMSC have resulted in identification of mesh reflectivity efficiencies for both gold plated molybdenum wire and copper plated dacron meshes. These tests indicate that gold plated wire mesh can be knitted to provide a reflectivity efficiency of 0.972 at 11 GHz and 0.987 at 5.1 GHz. This measured efficiency includes both the effects of energy transmitted through the material as well as the energy absorbed by the material. The dacron mesh flown on the ATS-6 vehicle exhibits a reflectivity efficiency of 0.9999 at 11 GHz. Since the highest possible efficiency is required and the thermal distortions, based on the analysis presented in Section 3.3.1, are insignificant with the dacron material, the dacron mesh should provide the best performance.

ORBIT	TIME	RMS FOR 5.1 GHz	η 5.1 GHz	RMS For 11 GHz	n 11 CHz
β = 0	0	0.777 x 10^{-3} λ	0.999904	0.746 x 10^{-3} λ	0.99991
	27	0.605 x 10^{-3} λ	0.99994	0.652 x 10^{-3} λ	0.99993
	54	0.346 x 10^{-3} λ	0.99998	0.373 x 10^{-3} λ	0.99998
	81	0.605 x 10^{-3} λ	0.99994	0.652 x 10^{-3} λ	0.99993
β = 30	0	0.648 x 10^{-3} λ	0.99993	0.652 x 10^{-3} λ	0.99993
	27	0.605 x 10^{-3} λ	0.99994	0.652 x 10^{-3} λ	0.99993
	54	0.346 x 10^{-3} λ	0.99998	0.373 x 10^{-3} λ	0.99998
	81	0.605 x 10^{-3} λ	0.99994	0.652 x 10^{-3} λ	0.99993
β = 80	0	0.346 x 10^{-3} λ	0.99998	0.373 x 10^{-3} λ	0.99998
	27	0.605 x 10^{-3} λ	0.99994	0.652 x 10^{-3} λ	0.99993
	54	0.346 x 10^{-3} λ	0.99998	0.373 x 10^{-3} λ	0.99998
	81	0.605 x 10^{-3} λ	0.99994	0.652 x 10^{-3} λ	0.99993

Figure 3.3-4 Effects of Orbit Thermal Distortion

For purposes of the efficiency calculation the 0.9999 will be used. This neglects the slight benefit which would be gained by adjusting the efficiency up to reflect the absorbed energy which will not contribute to the antenna main beam efficiency degradation.

3.3.3 Surface Approximation Effects

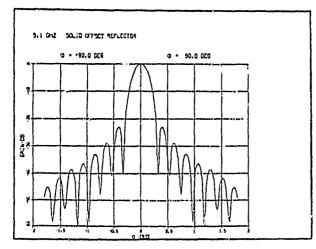
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The wrap rib surface approximation error has been well characterized. Figure 3.3-5 presents patterns of the perfect surface and the parabolic cylindrical gore approximation for the offset wrap rib at 5.1 GHz. The resulting main beam efficiency change was calculated to be 0.9708. The surface was also evaluated at 11 GHz. These results are presented in Figure 3.3-6. The efficiency effect of the approximation error was determined to be 0.9704 at the high frequency.

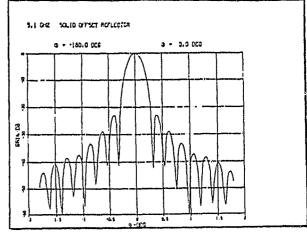
3.3.4 Manufacturing, Assembly and Deployment Effects

The effects related to our ability to produce a reflector surface approaching a zero error wrap rib approximation were evaluated. This evaluation was based on data developed during manufacturing, assembly and testing of the ATS-6 flight reflector and therefore should be reasonably representative of achievable tolerances.

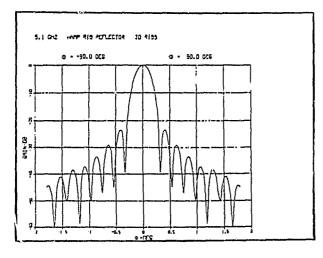
The significant manufacturing error is that of machining the parabolic contour on the rib. This operation if performed on a precision numerical control machine can be expected to achieve a final contour accuracy of 0.051 mm (0.002 inches) RMS. Upon reflector assembly each of the ribs must be mounted to the hub. Past experience indicates that the hinge and rib root can be adjusted within \pm 0.051 mm (0.002 inches) of the required position and the rib tips within \pm 0.254 mm (0.010 inches). During reflector deployment the rib locks up with some deployment repeatability error. This error has been measured at \pm 0.254 mm (0.010 inches) at the rib tip for reflectors with rib lengths of 8 to 10 m.

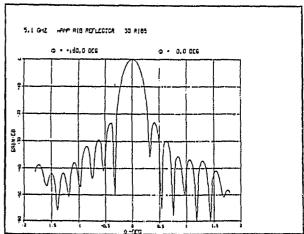


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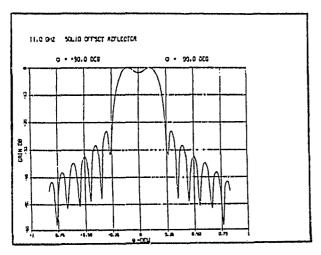
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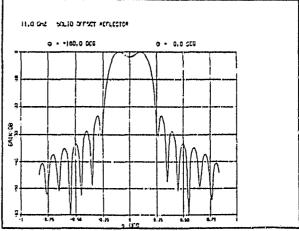




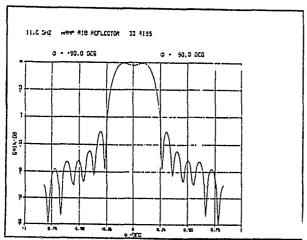
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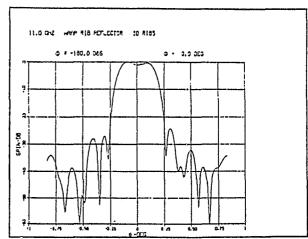
Figure 3.3-5 Reflector Pattern Predictions - 5.1 GHz





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Figure 3.3-6 Reflector Pattern Predictions - 11.0 GHz

Figure 3.3-7 presents the resulting surface errors and associated aperture efficiencies for the above effects. Since there is no truly accurate method for analytically distributing these errors and achieving a rigorous pattern, the main beam efficiencies are assumed to be equal to the more conservative aperture efficiencies.

3.3.5 Mechanical Scan Effects

The mechanical scan of 6 RPM produces a time invariable load on the reflector support structure and the reflector structure. This loading, being an invariant condition, can be pre-biased out as a source of surface error. However, in doing this it must be realized that spin rate variation must then be assessed as a source of error. In order to evaluate the spin rate variation effects, assumed at $\pm 1\%$ of the nominal 6 RPM rate and to define the initial bias shape requirements the reflector was analyzed for the spin induced distortions.

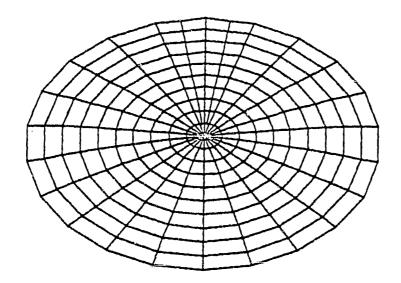
A NEPSAP, Nonlinear Elastic Plastic Structural Analysis Program, finite element model of the 15 M reflector was constructed. The model consisted of 241 modal points and 436 beam and membrane finite elements. This model is displayed in Figure 3.3-8. The ribs were represented as offset beam elements, the mesh as orthotropic membrane elements. The ribs were tapered lenticular elements with dimensions as shown below.

As can be seen from the table, two rib designs were evaluated:

	LOCATION	HEIGHT	WIDTH	THICKNESS
CASE 1	ROOT	177.8 mm	38.1 mm	0.508 mm
	TIP	50.8 mm	12.7 mm	0.508 mm
CASE 2	ROOT	254 mm	38.1 mm	0.508 mm
	TIP	50.8 mm	12.7 mm	0.508 mm

Figure 3.3-7
Manufacturing, Assembly and Deployment Effects

	RMS SURFACE E	RROR (ε)
SOURCE	5.1 GHz	11 GHz
RIB MANUFACTURING	$0.865 \times 10^{-3} \lambda$	1.318 \times 10 ⁻³ λ
RIB ROOT ASSEMBLY	$0.611 \times 10^{-3} \lambda$	$1.864 \times 10^{-3} \lambda$
RIB TIP ASSEMBLY	1.527 x 10^{-3} λ	1.208 x 10 ⁻³ λ
DEPLOYMENT	$1.527 \times 10^{-3} \lambda$	1.208 x 10 ⁻³ λ
TOTAL RMS SURFACE ERROR	$2.405 \times 10^{-3} \lambda$	2.847 x 10 ⁻³ λ
ASSOCIATED APERTURE EFFICIENCY*	0.9991	0.9987
$*_n = e^{-(4\pi\epsilon)^2}$		



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SOLID LINE UNDEFLECTED ORIGINAL FIGURE

Figure 3.3-8 Structural Model Overview

The orthotropic mesh stiffness properties employed in the model were based on ATS-6 mesh values and are defined as:

$$\begin{bmatrix} \text{Radial Line Load (N/mm)} \\ \text{Circumferential Line Load (N/mm)} \end{bmatrix} = \begin{bmatrix} 70 & 0 & 0 \\ 0 & 35 & 0 \\ 0 & 0 & 17.5 \end{bmatrix} \begin{bmatrix} \text{Radial Strain} \\ \text{Circumferential Strain} \\ \text{Shear Strain} \end{bmatrix}$$

To obtain the 6 RPM reflector distortions, a three increment non-linear analysis was performed. The first increment applied the circumferential mesh prestrain of 0.00088 N/mm (0.005 lb/in). This was then iterated to achieve convergence. The spin loads were then applied to produce the reflector distortions.

Figure 3.3-9 presents a summary of the rib tip deflections of the spin loaded reflector analyzed in Case 1. The developed surface RMS for the 6 RPM condition was calculated to be 30 mm (1.18 in) RMS. This uncompensated error would be unacceptable. Case 2 results, presented in Figure 3.3-10, yield a higher 32 mm (1.27 in) RMS. These results indicate substantial mesh effects due to spin which is consistent with the type of mesh employed. As a result the reflector must be biased for the nominal 6 RPM for weight and spin deflection considerations the lighter Case 1 rib was chosen for use in the design. A spin rate variation of + 1% of nominal was chosen for the error design requirement.

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The effects on efficiency ratioed from the 6 RMP case for a \pm 1% spin rate variation yield an 0.3 mm (0.0118 in) RMS surface for the 5.1 GHz case. This provides an aperture efficiency of 0.9959. The 11 GHz under illuminated case has an 0.11 mm (0.0043 in) surface RMS and a corresponding efficiency of 0.9975.

	LIBRARY 17
DISU NEP1 1 3 10/240/10	/DISU/NEP1/ 1/ 3/ NODAL DISPLACEMENT TABLE

1

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22	8.592E-06 7.354E-03 5.257E-03 5.257E-03 5.199E-03 7.354E-03 7.257E-03 7.357E-03 7.257E-03 7.257E-03 7.257E-03 7.257E-03 7.267E-03
>	3.013E-03- 7.559E-03- 7.174E-03- 1.357E-03- 2.475E-03- 3.352E-03- 3.365E-03- 3.365E-03- 3.365E-03- 7.26E-03- 7.26E-03- 7.
×	.160E-05 .879E-03 .75FE-03 .775E-03775E-03775E-03313E-04708E-03729E-0372
2	.366E+00 1 .374E+00-1 .736E+00-6 .776E+00-6 .410E+00-3 .402E-01-1 .402E-01-1 .402E-01-1 .338E+00 4 .632E+00 6 .760E+00 6 .760E+00 6 .760E+00 6 .760E+00 6 .760E+00 6 .760E+00 6 .760E+00 6
>	3.953E-03-1 3.953E-02-1 3.825E-02-1 3.92E-01-1 3.46E-01-1 3.46E-01-1 3.65E-01-1 3.64E-02-7 3.64E-02-7 3.78E-02-7 3.78E-01-1 3.78E-01-1 3.78E-01-1 3.78E-01-1 3.78E-01-1
×	1.213E+00 2 1.325E+00 7 1.443E+00 1 1.477E+00 1 1.103E+00 1 1.103E+00 1 1.103E+00 1 2.380E-01 1 6.748E-01 9 6.748E-01 9 6.748E-01 9 6.748E-01 1 1.102E+00-1 1.316E+00-1 1.473E+00-1 1.473E+00-1 1.473E+00-1 1.552E+00-1 1.552E+00-1
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Figure 3.3-9 6 RPM Rib Tip Deflections, 178 mm Rib

ORIGINAL CARRESTS OF POOR QUALITY

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DIO	>	88.74 9.3746 9.37776 9.37776 9.37776 9.37776 9.37776 9.377776 9.3777777777777777777777777777777777777
	×	.574E-06 .006E-03 .138E-03 .66EE-03 .657E-03 .180E-03 .197E-03 .1765E-04 .1765E-04 .1765E-04 .177E-03 .177E-03 .177E-03 .177E-03 .177E-03 .177E-03 .167E-03
LIBRARY 17	2	577E+00-2 578E+00-2 830E+00-5 950E+00-7 984E+00-5 143E+00-5 143E+00-5 906E+00-9 339E+00-9 1328E+00 2 965E+00 5 969E+00 5 378E+00 6 378E+00 5 377E+00 6 344E+00 6 344E+00 5 348E+00 5 348E+00 5 348E+00 5
LIBA	>-	2.441E-03-1. 3.985E-02-1. 1.120E-01-1. 1.398E-01-1. 1.594E-01-1. 1.560E-01-1. 1.560E-01-1. 1.362E-01-1. 1.331E-03-9. 5.943E-02-1. 1.381E-01-1. 1.381E-01-1. 1.365E-01-1. 1.572E-01-1. 1.572E-01-1. 1.573E-01-1. 1.573E-01-1. 1.573E-01-1. 1.573E-01-1. 1.573E-01-1. 1.573E-01-1. 1.573E-01-1. 1.573E-01-1.
/ 1/ 3/ LACEMENT	×	1.394E+00 1.386E+00 1.450E+00 1.551E+00 1.618E+00 1.618E+00 1.254E+00 1.010E+00 1.010E+00 1.010E+00 1.010E+00 1.010E+00 1.007E+00 1.252E+00 1.252E+00 1.252E+00 1.252E+00 1.252E+00 1.252E+00 1.252E+00 1.252E+00 1.252E+00 1.252E+00 1.252E+00
/DISU/NEP1. Nodal Dispi	INCR	
/DISU NODAL	NODE	00000000000000000000000000000000000000

Figure 3.3-10 6 RPM Rib Tip Deflections, 254 mm Rib

SU NEP1 1 3 10/240/10

APPENDIX A

SURFACE MODEL

The reflector surface model is being supplied in a NASTRAN compatible data file. This file contains the absolute coordinate locations of 186 reflector surface nodes equally distributed on each of the 30 ribs and mesh gores. Table A-1 presents a listing of the data file over half of the planar symmetric reflector surface. As presented in the table nodes 1 through 6 present surface points for rib 1, the rib which starts at the center of the offset and has a tip closest to the parent reflector vertex. Nodes 7 through 12 present the coordinates of the center of the mesh gore between rib 1 and 2. The sets of 6 nodes then continue to alternate rib and mesh gore nodes through rib 16. Thus displaying 16 ribs and 15 gores.

A right hand coordinate system is used with the origin located at the vertex of the parent parabola. The +Z axis points towards the focal point. The +Y axis is radial from the vertex passing through the offset reflector and is the axis of planar symmetry. All units in this table are in inches.

				5.4	.		
HODE NO.	4.3866			31	, 000	374.020	23,454
HINE ING.	K-COPD	Y-CORD	2-00PD	35	60,233	260,661	20, 435
				33	93, 483	213.553	15, 376
_				34	114,264	177, 356	12,019
i	. 000	374.020	39,434	35	131.711	146.797	11.052
2 3 4	.000	241.976	16.490	36	147.032	119.344	10.245
3	.000	187.287	9. 328	**	*		1015.13
4	.000	145.323	5.353				
<u> </u>	.000	109.948	3.270	37	.000	374.020	39, 494
5 6	.000	73.732		á á	73.243	267.657	21.913
•	.000	101105	1.575				
				33	110.379	223.393	17.454
7	344			40	134.923	139,353	15.153
ŗ	.000	374.020	33.484	41	155.539	160,607	13, 309
a a	13.913	243.459	16.803	42	173,645	135.236	13.479
	19.616	189.370	10.263				
10	23,968	147.363	6.390				
11	27.620	112.869	3.337	43	,000	374.020	33,434
12	30.326	82.036	2.262	4.4	33.419	276.917	23.363
			47442	45	124,757	236.434	20.207
				46	152.527	205.266	18.513
13	.000	374.020	39.484	47	175.860	179.922	17.337
14	27.327	244,942	17.116	49			
15	39.232			+0	196.359	155.659	17,312
		191,454	10.703				
16	47,936	150.403	5.926				
17	55.241	115.789	4.504	49	.000	374.020	39.484
18	61.652	35.290	2.950	50	93.539	286.177	25.324
				51	139.135	249.469	22.960
				52	170.130	221.175	21.373
19	.000	374.020	39.484	53	196.192	197.336	21.705
20	41.072	249.303	19.037	54	219.074	176.091	22.125
21	57.914	197.536	12.003	**		., .,	
22	70.770	157.878	3,505				
23	31.562	124.388	5.320	55	.000	374.020	39.484
24						297.251	29.162
24	91.035	94,371	4.973	56	105.568		
				57	150.437	265.074	26.255
				53	183.791	240.233	25.997
.35	.000	374.020	39.494	59	212.204	219.191	26.341
26	54.317	253,665	13.959	60	237.004	200.576	27.298
27	76.596	203.718	13.279				
23	93,604	165.354	10.034				
29	107.833	132.397	3.136	61	.000	374.020	39.484
30	120.419	104.452 .	5.997	53	114,546	308.326	30.500
			*****	63	161.739	290.678	29.550
				54	197.851	259.291	29.922
				54 65	228.225	241.145	30.977
				ခ်င်	254,933	225.070	32,471

Table A-1 Reflector Surface Coordinates

67 69 70 71 72	.000 120.050 169.564 207.471 239.370 267.429	374.020 320.721 298.165 290.686 265.786 252.580	39.484 33.119 33.242 34.435 36.130 33.279	97 99 99 100 101 102	,000 131,456 185,961 227,304 263,092 294,191	374.020 336.551 391.168 394.483 397.124 399.335	39,494 47,017 52,379 53,469 63,911 69,264
73 74 75 76 77	.000 125.554 177.389 217.091 250.514 279.925	374.020 333.117 313.651 303.041 290.427 280.090	39,434 35,735 36,934 38,949 41,394 44,033	103 104 105 106 107 103	,000 128.792 192.249 223,309 257,952 288,494	374,020 399,340 409,996 417,552 423,793 429,160	39.494 49.822 56.335 69.536 75.560
79 30 31 32 33 34	.000 128.404 191.472 223.144 256.398 286.552	374.020 346.326 334.307 324.867 316.762 309.510	39.484 39.524 40.873 43.768 46.944 50.300	109 110 111 112 113 114	.000 126.129 173.536 219.314 252.312 282.797	374.020 413.130 429.803 440.620 450.441 458.986	39.434 52.628 60.324 63.206 75.166 81.857
35 35 37 33 39	,000 131,253 185,556 227,196 262,282 293,179	374.020 359.534 352.962 347.692 343.096 338.931	39,484 41,313 44,312 49,597 52,504 56,512	115 116 117 113 119 120	.000 120.763 170.990 209.610 242.223 270.995	374.020 425.675 446.598 462.443 475.680 487.243	39.494 55.276 64.579 72.812 30.493 87.821
51 93 94 95 96	.000 131.355 135.758 227.500 262.687 293.685	374.020 373.043 372.065 371.083 370.110 369.133	39.494 44.165 49.846 53.526 53.207 62.883	131 133 134 136 126	. 000 115,358 153,443 200,407 231,533 258,133	374,020 433,820 464,373 434,266 500,313 515,499	39.934 57.925 63.333 77.419 35.321 93.785
				127 129 129 130 131 132	.000 107.497 152.276 196.749 215.381 241.601	374.020 449.439 430.366 503.907 523.649 540.965	39.434 60.304 71.709 31.565 90.619 99.160

Table A-1 Reflector Surface Coordinates (Continued)

133	.000	374.020	39,494
134	99.576	460,757	52,652
135	141.109	496.359	75,055
136	173.091	523.549	55,710
137	300.129	546.331	95,416
133	224.008	566.430	104,535
139	,000	374.020	39.484
140	99,339	470.224	84.881
141	126,696	509.308	77.923
142	155,435	540.076	39.199
143	179,739	565.522	99.456
144	201,207	597.886	109.063
145 146 147 143 149	.000 79.200 112.294 137.790 159.349 179.406	374.020 479.691 523.257 556.605 594.663 609.342	39,484 56,679 30,762 92,633 103,496 113,591
151	,000	374.020	39.484
152	67.134	486.870	68.195
153	95.190	533.463	32.916
154	116.317	569.156	35.337
155	135.116	599.206	106.565
156	151.287	625.649	117.032
157	.000	374.020	39, 484
158	55.067	494.049	69, 710
159	73.096	543.670	65, 071
160	95.354	591.707	97, 985
161	110.383	613.748	109, 634
162	124.169	641.956	120, 474
163	.000	374.020	39,484
164	41.654	499,536	70,657
165	59.077	550.053	86,418
166	72.514	589.560	99,643
167	93:887	622.950	111,555
168	93.942	652.167	122,628
169	.000	374.020	39.484
170	28.240	503,034	71.604
171	40.053	55%.437	97.765
172	49.174	597.413	101.300
173	56.392	631.952	113.476
174	63.716	662.377	124.793
175	.000	374.020	39.484
176	14.120	504.551	71.927
177	20.029	558.610	83.224
178	34.537	600.038	101.865
179	28.446	635.053	114.130
180	31.858	665.855	125.517
131 182 183 184 185 186	.000 .000 .000 .000	374.020 506.079 560,734 602.762 638.153 669.334	39.484 72.249 88.683 102.429 114.784 126.251

Table A-1 Reflector Surface Coordinates (Continued)